International Application No.: PCT/JP2004/016434

U.S. Patent Application No.: Unknown

March 14, 2005 Page 10 of 10

REMARKS

Claims 7-18 are pending in this application. By this Preliminary Amendment, Applicant AMENDS the specification and the abstract of the disclosure, CANCEL claims 1-6 and ADDS new claims 7-18.

Applicant has attached hereto a Substitute Specification in order to make corrections of minor informalities contained in the originally filed specification. Applicant's undersigned representative hereby declares and states that the Substitute Specification filed concurrently herewith does not add any new matter whatsoever to the above-identified patent application. Accordingly, entry and consideration of the Substitute Specification are respectfully requested.

The changes to the specification have been made to correct minor informalities to facilitate examination of the present application.

Applicant respectfully submits that this application is in condition for allowance. Favorable consideration and prompt allowance are respectfully solicited.

Respectfully submitted,

Date: March 14, 2005

Attorneys for Applicants
Joseph R. Keating
Registration No. 27 269

Registration No. 37,368

Christopher A. Bennett Registration No. 46,710

Peter Medley Registration No. 56,125

KEATING & BENNETT, LLP 10400 Eaton Place, Suite 312

Fairfax, Virginia 22030 Telephone: (703) 385-5200

Facsimile: (703) 385-5080



DT09 Rec'd PCT/PT0 1 4 MAR 2005

DESCRIPTION

Attorney Docket No. 36856.1330

ELECTROMAGNETIC FIELD ANALYZER, ELECTROMAGNETIC FIELD ANALYZING PROGRAM, AND STORAGE MEDIUM FOR RECORDING THE PROGRAM

TECHNICAL BACKGROUND OF THE INVENTION

1. Field of the Invention

The present invention relates to electromagnetic field simulation technology used in development developing and designing of high-frequency products such as multilayer products, giga filters, and EMI (Electro-Magnetic Interference) elimination filters, etc., and more particularly, the present invention relates to an electromagnetic field analyzer, an electromagnetic field analyzing program, and a storage medium for recording the program in which the simulation time can be shortened is reduced.

Background

2. Description of the Related Art

In recent years, the development and designing of high-frequency products have been vigorously performed aggressively pursued, and, in order to perform an electromagnetic field analysis of high-frequency products, electromagnetic field simulation softwares have been widelysoftware has been commonly used. Generally, there are many eases instances where a finite element method is used for electromagnetic field analysis. In the electromagnetic field simulation using the finite element method, since it takes a long time to solve simultaneous linear equations, high-speed solving methods are

desired. In particular, in the field of electromagnetic field analysis, since nothing except for only a direct method called a Gauss' elimination method can be used, the problem of taking a long calculation time—times is more serious.

However, by a paper (R. Hiptmair, "Multigrid methodMethod for Maxwell's equations Equations," SIAM Journal of Numerical Analysis, vol. 36, no. 1, pp. 204-225, 1999+), which is referred to herein as Non-patent document 1, and which was made public by Hiptmair in 1999, it is proved that a solution method called a multi-grid method can be also used in an electromagnetic field analysis and the method is paid attention to as a greatly speeding up solution when reduces the calculation time as compared with—to a direct method.

In the multi-grid method, an analysis object is divided into two classes of fine elements and coarse elements, and the solution of fine elements is found by making use of determined using the solution of coarse elements.

When simultaneous linear equations are calculated by using the Gauss' elimination method, generally the calculation time is proportional to the third power of the dimensions of a matrix. Since the size of the dimensions corresponds to the number of elements, when the number of elements becomes double doubled, the calculation time becomes—is 8 times as large, and, when the number of elements becomes—is 10 times as large, the calculation time becomes 1,000 times as large.

Then, when—When the Gauss' direct method is applied to only coarse elements, an approximate solution to fine elements is obtained by using a—the solution to coarse elements, and the Gauss' elimination method is not used, the calculation time

for fine elements can be reduced to 1/8 and 1/1000, respectively. Since additional calculation time is added, although the practical calculation time is not reduced as—to such a great extent, the calculation time can—be is always greatly reduced—without—fail.

The multi-grid method is classified can be broken down into two methods: a method using a nested mesh and a method using a non-nested mesh. However, since a prolongation matrix is incomplete as described later, there is nothing except for using a the method using a nested mesh must be used.

Figs. 12A and 12B are drawings for describingshow the division into elements using a nested mesh. Fig. 12A shows the division into elements using a nested mesh in the case of two-dimensional analysis where a coarse triangular element is uniformly divided to form fine triangular elements.

Furthermore, Fig. 12B shows the division into elements using a nested mesh in the case of three-dimensional analysis where a coarse tetrahedral element is uniformly divided to form eight fine tetrahedral elements. As is—understood from Figs. 12A and 12B, there are geometrical restrictions between coarse elements and fine elements in—the case of when using a nested mesh.

Non-Another prior art method is described in non-patent document 1: R. Hiptmair, "2: D. Dibben and T. Yamada, "Non-Nested Multigrid method for Maxwell's equations," SIAM Journal of Numerical Analysis, vol. 36, No. 1, pp. 204 225, 1999

Non-patent document 2: D. Dibben and T. Yamada, "Nonnested multigrid and automatic mesh coarsening for high
frequency electromagnetic problems and Automatic Mesh

Coarsening for High Frequency Electromagnetic Problems" IEEJ Investigating Research Committee material Material, SA-02-34, pp. 71-75, 2002.

Figs. 13A and 13B are drawings for describingdescribe a first problem of that occurs when using a nested mesh. Fig. 13A shows how a circle is divided into elements using a nested mesh to form coarse elements. Furthermore, Fig. 13B shows how coarse elements are divided into fine elements using a nested mesh. As is—understood from Figs. 13A and 13B, a curved surface cannot be correctly expressed because of the geometrical restrictions of a nested mesh.

Furthermore, Fig. 13C shows how a circle is divided into elements using a non-nested mesh to form fine elements. As is understood from Fig. 13C, since there are no geometrical restrictions when using a non-nested mesh, a curved surface can be correctly expressed.

Figs. 14A to 14C are to describe a second problem of that occurs when using a nested mesh. Fig. 14A shows how a square is divided into elements by using a nested mesh to form coarse elements. Furthermore, Fig. 14B shows how coarse elements are divided into fine elements. As is understood from Figs. 14A and 14B, coarse elements cannot be partially divided into fine elements because of the geometrical restrictions of a nested mesh.

Furthermore, Fig. 14C shows how a square is divided into elements by—using a non-nested mesh to form fine elements. As is—understood from Fig. 14C, coarse elements can be partially divided into fine elements because of there are no geometrical restrictions when using a non-nested mesh.

Most industrial products have curved surfaces, such as in a circular cylinder and a—square <u>surfaces</u>, and they inevitably contain partially fine portions. Accordingly, it is difficult to apply a nested mesh to industrial products. On the other hand, since there <u>is—are</u> no geometrical restrictions in a nonnested mesh, fine elements can be freely formed. However, since the positional <u>relationrelationship</u> is not systematic, it is difficult to make the electromagnetic field of coarse elements related to the electromagnetic field of fine elements. It is a prolongation matrix to make coarse elements related to fine elements, but no precise prolongation matrix has been found. This means that a multi-grid method cannot be practically used, and accordingly, nothing can be used except for a Gauss' direct method, which takes much—requires a lot of time, in the analysis of magnetic field.

Disclosure of Invention

The present invention has been done to-solve the SUMMARY OF THE INVENTION

To overcome the problems described above problem, and it is an object, preferred embodiments of the present invention to provide an electromagnetic field analyzer, an electromagnetic field analyzing program, and a recording medium having the program recorded thereinthereon, in which the analysis of an electromagnetic field can be made by using performed via a multi-grid method using a non-nested mesh—

It is another object of the present invention to provide—, and an electromagnetic field analyzer, an electromagnetic field analyzing program, and a recording medium in which the program has been recorded, in which the time required for an

electromagnetic field analysis is greatly reduced.

According to an aspecta preferred embodiment of the present invention, an electromagnetic field analyzer comprises includes dividing means for dividing form data as an analysis object into coarse elements and fine elements; forming means for forming a matrix makingdefining an electromagnetic field vector of the coarse elements divided by the dividing means related to an electromagnetic field vector of the fine elements, and calculating means for calculating an approximate solution of the electromagnetic field vector of the fine elements by applying an iteration method of simultaneous linear equations while referring to the matrix formed by the forming means.

Preferably, the forming means forms the matrix by expressing the elements of electromagnetic field vectors at the sides at a fine element using an interpolation function in the coarse elements.

More preferably, when the length of a side l_i of the fine element is $|l_i|$, an interpolation function showing the relationrelationship between the electromagnetic field at position x of the side l_i of the fine element and the electromagnetic field at a side j of the coarse element is N_j^c (x), and the tangential vector of the side l_i of the fine element is t_i , the forming means (25)—forms a matrix P_{ij} preferably by Expression (14) to be described later.

More preferably, in the <u>calculation</u>—<u>calculating</u> means, the accuracy of an approximate solution of the electromagnetic field vector of the fine elements is improved such that high-frequency components included in the approximate solution of

the electromagnetic field vector of the fine element are removed by applying a stationary iteration method of simultaneous linear equations, <u>such</u> that a residual in the fine elements is mapped to a residual in the coarse elements by using the matrix formed by the forming means, <u>such</u> that a correction vector to the coarse elements is formed by applying a direct method or a non-stationary iteration method of simultaneous linear equations, and <u>such</u> that a correction vector to the fine elements is obtained by using the matrix formed by the forming means.

According to another aspectpreferred embodiment of the present invention, in a computer program to causecontrol a computer to execute an electromagnetic field analyzing method for analyzing an electromagnetic field as an analysis object, the computer contains includes a first storage means for storingdevice arranged to store divided elements and a second storage means for storingdevice arranged to store a matrix, and the electromagnetic field analyzing method contains includes the steps of dividing form data as an analysis object into coarse elements and fine elements and storing the elements in the first storage means device, forming a matrix makingdefining an electromagnetic field vector of the coarse elements stored in the first storage means device related to an electromagnetic field vector of the fine elements and storing the matrix in the second storage means device, and calculating an approximate solution of the electromagnetic field vector of the fine elements by applying an iteration method of simultaneous linear equations while referring to the matrix stored in the second storage

meansdevice.

According to another aspectpreferred embodiment of the present invention, in a computer readable storage medium, being computer readable, for recording having stored thereon a program to causecontrol a computer to execute an electromagnetic field analyzing method for analyzing an electromagnetic field as an analyzing object, the computer contains includes a first storage means for storing device arranged to store divided elements and a second storage means for storing device arranged to store a matrix, and the electromagnetic field analyzing method containsincludes the steps of dividing form data as an analysis object into coarse elements and fine elements and storing the coarse and fine elements in the first storage meansdevice, forming a matrix makingdefining an electromagnetic field vector of the coarse elements stored in the first storage means device related to an electromagnetic field vector of the fine elements and storing the matrix in the second storage means device, and calculating an approximate solution of the electromagnetic field vector of the fine elements by applying an iteration method of simultaneous linear equations while referring to the matrix stored in the second storage meansdevice.

Since the forming means forms a matrix makingdefining an electromagnetic field vector of coarse elements divided by the dividing means related to an electromagnetic field vector of fine elements, it became—is possible to perform an electromagnetic field analysis by—usingvia a multi-grid method using a non-nested mesh. Furthermore, since the calculating

means calculates an approximate solution of the electromagnetic field vector of fine elements while referring to the matrix formed by the forming means, it the calculating means is not required to calculate the electromagnetic field of fine elements by using a direct method, it became possible to largely reduce a. Thus, the time neededrequired for an electromagnetic field analysis is greatly reduced.

Furthermore, since the forming means forms a matrix such that an electromagnetic field vector element at a side of fine elements is expressed by using an interpolation function in coarse elements, it became possible to a matrix is easily form a matrixformed.

Furthermore, in the forming means, since a matrix P_{ij} is formed <u>preferably</u> by Expression (14) to be described later, a matrix <u>ean beis even</u> more easily formed.

Furthermore, in the calculating means, an approximate solution of an electromagnetic field vector of fine elements is corrected by using a matrix formed by the forming means, it became possible to improve. Thus, the accuracy of the approximate solution of an electromagnetic field vector of fine elements is greatly improved at high speed.

Brief Description of the Drawings

Other features, elements, steps, characteristics and advantages of the present invention will become more apparent from the following detailed description of preferred embodiments thereof with reference to the attached drawings.

BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 is an external appearance of shows an

electromagnetic field analyzer according to analyzer according to analyzer embodiment of the present invention.

- Fig. 2 is a block diagram showing the construction of the electromagnetic field analyzer according to analyzered embodiment of the present invention.
- Fig. 3 is a block diagram showing the functional structure of the electromagnetic field analyzer.
- Fig. 4 is a flow chart for describing the procedure of the electromagnetic field analyzer.
- Fig. 5A shows a microstrip line divided into large elements.
- Fig. 5B shows a microstrip line which is analyzed using the large elements shown in Fig. 5A, and in which the elements are subdivided based on their error estimation.
- Fig. 6 shows components of an electromagnetic field which that are in contact with the sides of a tetrahedron.
- Fig. 7 shows the relation relationship between an electromagnetic field vector \mathbf{u}^F in Ω^F and an electromagnetic field vector \mathbf{u}^C in Ω^C .
- Fig. 8 shows one example of a rectangular waveguide as an analysis object of an electromagnetic field analyzer according to $\frac{\partial}{\partial x}$ preferred embodiment of the present invention.
- Fig. 9 shows the calculation time to the number of elements when the rectangular waveguide shown in Fig. 8 is driven in TE10 mode of 2.45 GHz, which is compared between the case where a related direct method is used and the case where a method of a preferred embodiment of the present invention is used.

Fig. 10 shows one example of a patch antenna as an analysis object of the electromagnetic field analyzer according to analyzerdembodiment of the present invention.

Fig. 11 shows the calculation time to the number of elements when the patch antenna shown in Fig. 10 is driven at 7.0 GHz, which is compared between the case where a related direct method is used and the case where a method of the present invention is used.

Fig. 12A shows elements divided by using a nested mesh in the case of two-dimensional analysis.

Fig. 12B shows elements divided by using a nested mesh in the case of three-dimensional analysis.

Fig. 13A shows coarse elements formed such that a circle is divided into elements by using a nested mesh.

Fig. 13B shows fine elements into which coarse elements are divided by using a nested mesh.

Fig. 13C shows fine elements formed such that a circle is divided into elements by using a non-nested mesh.

Fig. 14A shows coarse elements formed such that a square is divided into elements using a nested mesh.

Fig. 14B shows fine elements into which coarse elements are divided by using a nested mesh.

Fig. 14C shows fine elements formed such that a square is divided into elements by using a non-nested mesh.

Reference Numerals

DETAILED DESCRIPTION OF PREFERRED EMBODIMENTS

Fig. 1——computer main body

- 2 display device ---3 FD-drive ----4 FD -- 5 keyboard - 6 mouse - 7 CD ROM-device 8 CD ROM - 9 - network communications device -- 10 CPU 12 RAM 13 hard disk 21 form data storage portion 22 division into element portion - 23 element storage portion - 24 - vector/matrix forming portion 25 prolongation matrix forming portion 26 vector/matrix-storage portion 27 prolongation matrix storage portion ----- 28 -- approximate value calculation portion 29 approximate solution correction portion ----30 -- electromagnetic field vector storage portion Best Mode for Carrying Out the Invention -- Fig. 1 is an external appearance of shows an electromagnetic field analyzer according to analyzer according to embodiment of the present invention. The electromagnetic field analyzer containspreferably includes a computer main body 1, a display device 2, an FD (Flexible Disk) drive 3 to

which an FD 4 is attached, a keyboard 5, a mouse 6, a CD-ROM

(Compact Disk- Read Only Memory) drive 7 to which a CD-ROM 8 is attached, and a network communications device 9. An electromagnetic field analyzing program is supplied through a recording medium such as an FD 4, or a CD-ROM 8, etc.or other suitable media. An electromagnetic field analysis is performed such that the electromagnetic field analyzing program is executed by the computer main body 1. Furthermore, the electromagnetic field analyzing program may be supplied to the computer main body 1 by antheranother computer through the network communications device 9.

Fig. 2 is a bock diagram showing the construction of the electromagnetic field analyzer of-anaccording to a preferred embodiment of the present invention. The computer main body 1 shown in Fig. 1 contains 1 includes a CPU (Central Processing Unit) 10, a ROM (Read Only Memory) 11, a RAM (Random Access Memory) 12, and a hard disk 13. The CPU 10 conductsperforms processing while the CPU 10 inputs and outputs data to the display device 2, the FD drive 3, the keyboard 5, the mouse 6, the CD-ROM device 7, the network communications device 9, the ROM 11, the RAM 12, or the hard disk 13. The electromagnetic field analyzing program in the FD 4 or the CD-ROM 8 is preferably stored inon the hard disk 13 through the FD drive 3 or the CD-ROM drive 7. The electromagnetic field analysis is performed such that the CPU 10 properly loads the electromagnetic field analyzing program from the hard disk 13 in the RAM 12 and executes the program.

Fig. 3 is a block diagram showing the functional structure of the electromagnetic field analyzer according to $\frac{1}{2}$ preferred embodiment of the present invention. The

electromagnetic field analyzer contains includes a form data storage portion 21 for storing form data as an analysis object, a division-into-element portion 22 for dividing the form data_ such that the size of elements is changed, an element storage portion 23 for storing the elements divided by the divisioninto-element portion 22, a vector/matrix forming portion 24 for forming a vector and matrix in simultaneous linear equations, a prolongation matrix forming portion 25 for forming a prolongation matrix, a vector/matrix storage portion 26 for storing the vector and matrix formed by the vector/matrix forming portion 24, a prolongation matrix storage portion 27 for storing the prolongation matrix formed by the prolongation matrix forming portion 25, an approximate value calculation portion 28 for calculating an approximate value of an electromagnetic field vector, an approximate solution correction portion 29 for correcting the approximate solution calculated by the approximate value calculation portion 28, and an electromagnetic-field vector storage portion 30 for storing, as an electromagnetic field vector, the approximate solution after having been corrected by the approximate solution correction portion 29.

Moreover, the form data storage portion 21, the element storage portion 23, the vector/matrix storage portion 26, the prolongation matrix storage portion 27, and the electromagnetic field vector storage portion 30 are provided in a fixed area in the RAM 12 or on the hard disk 13 shown in Fig. 2. Furthermore, the <u>functionfunctions</u> of the division-into-element portion 22, the vector/matrix forming portion 24, the prolongation matrix forming portion 25, the approximate

value calculation portion 28, and the approximate solution correction portion 29 is realizedare arranged such that the CPU 10 shown in Fig. 2 executes the electromagnetic field analyzing program loaded into the RAM 12.

Fig. 4 is a flow chart for describing the procedure of the electromagnetic field analyzer according to analyzer ampreferred embodiment of the present invention. First, the division-into-element portion 22 reads form data as an analysis object stored in the form data storage portion 21, divides the form data into elements, and preserves the divided elements in the element storage portion 23 (S11). The division-into-element portion 22 divides the form data into triangular or quadrangular elements in the case of two-dimensional analysis, and divides the form data into tetrahedral, triangle-pole-shaped, or square-pole-shaped elements in the case of three-dimensional analysis.

The division-into-element portion 22 divides the form data while changing the size of the elements. In order to make the description simple, here—elements of two different sizes are formed and it is assumed that a set of larger (coarse) elements is made Ω^c and a set of smaller (fine) elements is made Ω^F . Moreover, a superscript of c represents coarse, and a superscription of F represents fine.

The sets of Ω^c and Ω^F may be independently formed, buthowever, when the set of Ω^c is first formed and then the set of Ω^F is formed by subdividing the elements, the processing time is shortened.reduced. Regarding the method of subdivision, there are is a method for subdividing elements which are larger in volume and a method for analyzing elements

by using Ω^c and subdividing elements having <u>a</u> large error. Either of them may be used, <u>buthowever</u>, the method in which error is evaluated has <u>a</u>—better calculation accuracy. Moreover, regarding the method for evaluating error, <u>Reference</u> Document No. 1 (Norio Kamiya et al., <u>"computer and application's monk, special edition/error estimation in software and adaptive element," No. "Computer and Application's Monk, Special Edition/Error Estimation in Software and Adaptive Element," No. 42, CORONA PUBLISHING CO., 1993)) can)</u>

Figs. 5A and 5B show examples of division into elements in a microstrip line. In Fig. 5A, the microstrip line is divided into larger elements. Furthermore, in Fig. 5B, an analysis is performed by using the larger elements in Fig. 5A and the elements are subdivided based on their error estimation.

Since the electromagnetic field is concentrated at the edge of an electrode, the portion is subdivided into smaller elements.

Next, the vector/matrix forming portion 24 reads the sets Ω^c and Ω^F stored in the element storage portion 23, discretization of the sets takes place using their elements such that a finite element method is applied to a Maxwell equation, and a matrix and column vector in simultaneous linear equations are formed (S12). Here, the Maxwell equation as an object is represented by the following expression—s:

Expression 1 $\nabla \times H = (\hat{\sigma} + j\omega\hat{\epsilon})E \qquad \cdots (1)$ $\nabla \times E = -j\omega\hat{\mu}H \qquad \cdots (2)$ $\frac{\text{Here}_{\tau}}{\nabla \times H} = (\hat{\sigma} + j\omega\hat{\epsilon})E \qquad \cdots (1)$ $\nabla \times E = -j\omega\hat{\mu}H \qquad \cdots (2)$

Where E: electric field, H: magnetic field, ϵ : dielectric constant, μ : permeability, σ : conductivity, ω : angular frequency, ∇ : differential operator $(\partial/\partial x, \partial/\partial y, \partial/\partial z)$, X: outer product, j: imaginary number

From these expressions, the following differential equation can-beis obtained:

Expression
$$\frac{23}{\sqrt{3}}$$

$$\nabla \times \hat{p} \nabla \times u - \omega^2 \hat{q} u = 0 \qquad \cdots (3)$$

u represents an electric field, magnetic field, etc., and \hat{p} and \hat{q} represent characteristic values of dielectric constant and permeability. For example, when Expression (3) is obtained by substituting Expression (2) into Expression (1) and eliminating H, u represents an electric field, \hat{p} represent the reciprocal of a permeability, and \hat{q} represents a dielectric constant.

When a finite element method is applied to an element V_k formed in the above-former process described above, the following expression ean be-is obtained:

Expression 3

$$\int_{V_{k}} (\nabla \times N_{ki}) \hat{p}(\nabla \times u) dV - \omega^{2} \int_{V_{k}} N_{ki} \hat{q} u dV = \int_{\partial V_{k}} (N_{ki} \times \hat{p} \nabla \times u) \cdot n dS \cdots (4)$$

Here, ∂V_* represents the surface of V_* and n represents an outward normal unit vector. N_{ki} represents an interpolation function (also called a shape function, insertion function, basis function, etc.) When tetrahedral elements and linear equations are used, the interpolation function is given by the following expression.

Expression 4

$$\int_{V_k} (\nabla \times N_{ki}) \hat{p}(\nabla \times u) dV - \omega^2 \int_{V_k} N_{ki} \hat{q} u dV = \int_{\partial V_k} (N_{ki} \times \hat{p} \nabla \times u) \cdot n dS$$

$$\cdots (4)$$

Here, ∂V_k represents the surface of V_k and n represents an outward normal unit vector. N_{ki} represents an interpolation function (also called a shape function, insertion function, or basis function). When tetrahedral elements and linear equations are used, the interpolation function is given by the following expression:

Expression 5
$$N_{k1} = |l_1|(L_2\nabla L_3 - L_3\nabla L_2)$$

$$N_{k2} = |l_2|(L_3\nabla L_1 - L_1\nabla L_3)$$

$$N_{k3} = |l_3|(L_1\nabla L_2 - L_2\nabla L_1)$$

$$N_{k4} = |l_4|(L_1\nabla L_4 - L_4\nabla L_1)$$

$$N_{k5} = |l_5|(L_2\nabla L_4 - L_4\nabla L_2)$$

$$N_{k6} = |l_6|(L_3\nabla L_4 - L_4\nabla L_3)$$
... (5)

Here, L_1 to L_4 represent a volume coordinate system and $\left| l_1 \right|$ to $\left| l_6 \right|$ represent the length of a side l_i . Moreover, regarding the detail of the volume coordinate system, Reference Document 2 (T. Itoh, G. Pelosi and P. P. Silvester, "Finite Element Software for Microwave Engineering," John Wiley & Sons, pp. 101-125, 1996) canmay be referred to.

Fig. 6 shows components of an electromagnetic field u which that are in contact with the sides of a tetrahedron. The components of the electromagnetic field u are disposed arranged so as to be disposed along the six sides, and they are represented by u_1 to u_6 . The electromagnetic field at

point x inside the element V_k can be is represented by the following expression using Expression (56) and the six components of an electromagnetic field—:

--- Expression 5

$$-u(x) = \sum_{i=1}^{6} N_{ki}(x)u_{ki}$$
 ... (6)

When Expression (6) is substituted into Expression (4) and applied to all the elements of $\Omega^{\rm F}$, the following simultaneous linear equations can be obtained.

Expression 6

$$u(x) = \sum_{i=1}^{6} N_{ki}(x)u_{ki}$$
 ... (6)

Here, u^F is vectors which are electromagnetic fields disposed at the sides, and the dimensions are equal to the number of the sides as a general rule. The components of a matrix A^F are formed by the following expression:

When Expression (6) is substituted into Expression (4) and applied to all of the elements of $\Omega^{\rm F}$, the following simultaneous linear equations are obtained:

Expression 7

$$A^F u^F = b^F \qquad \cdots (7)$$

Here, u^F represents vectors which are electromagnetic fields disposed at the sides, and the dimensions are equal to the number of the sides as a general rule. The components of a matrix A^F are formed by the following expression:

Expression 8

$$\sum_{k=1}^{N^F} \sum_{i=1}^{6} \sum_{j=1}^{6} \left[\int_{V_k} (\nabla \times N_{ki}) \, \hat{p} (\nabla \times N_{kj}) \, dV - \omega^2 \int_{V_k} N_{ki} \hat{q} N_{kj} \, dV \right] \qquad \cdots (8)$$

Furthermore, a vector b^F is formed by the following

expression.

Expression 89

$$\sum_{k=1}^{N^F} \sum_{i=1}^{6} \sum_{j=1}^{6} \int_{\partial V_k} (N_{ki} \times \hat{p} \nabla \times N_{kj}) \cdot ndS \qquad \cdots (9)$$

Here, N^F is the number of elements in Ω^F .

The vector/matrix forming portion 24 also forms a matrix A^{C} corresponding to Ω^{C} invia the same procedure and preserves A^{F} , b^{F} , and A^{C} in the vector/matrix storage portion 26. When u^{F} is founddetermined by calculating simultaneous linear equations in Expression (7), the electromagnetic field distribution can be calculated using Expression (6) to result inproduce the solutions of Maxwell equations in Expressions (1) and (2).

Next, the prolongation matrix forming portion 25 reads elements Ω^F and Ω^C formed in the former process from the element storage portion 23 and, based on the geometric information, the prolongation matrix forming portion 25 forms a prolongation matrix P which makes relates the electromagnetic field vector \mathbf{u}^F in Ω^F related to the electromagnetic field vector \mathbf{u}^C in Ω^C (S13). When the prolongation matrix is used, as shown in the following expressions, one vector can makebe used to determine the other vector—determined:

- Expression 9

Expressions 10 and 11
$$u^{F} = Pu^{C} \qquad \cdots (10)$$

$$u^{C} = P^{t}u^{F} \qquad \cdots (11)$$

Here, t means a transposed matrix.

Fig. 7 shows the relationrelationship between an

electromagnetic field vector u^F in Ω^F and an electromagnetic field vector u^C in Ω^C . The j component u_j^C of the column vector u^C is disposed at the side of an element of Ω^C and the electromagnetic field u(x) at the position x in Ω^F can be represented using an interpolation function N^C in Ω^C .

---Expression 10

$$-u(x) = \sum_{j} N_{j}^{C}(x)u_{j}^{C} \qquad \cdots (12)$$

The i component u_i^F of u^F is made an average value of an electromagnetic field at the side l_i where the i component u_i^F is disposed. That is, the i component u_i^F of u^F is represented by the following expression.

- Expression 11

$$u_i^F = \frac{1}{|l_i|} \int_{l_i} u(x) \cdot t_i dl \qquad \cdots (13)$$

Here, t_i represents a tangential unit vector of the side l_i . When Expression (12) is substituted for Expression (13) and compared with Expression (10), the ij component P_{ij} of a prolongation matrix P is given by the following expression.

Expression 12

$$u(x) = \sum_{j} N_{j}^{C}(x)u_{j}^{C} \qquad \cdots (12)$$

The i component u_i^F of u^F is an average value of an electromagnetic field at the side l_i where the i component u_i^E is disposed. That is, the i component u_i^F of u^F is represented by the following expression:

Expression 13

$$u_i^F = \frac{1}{|l_i|} \int_{l_i} u(x) \cdot t_i dl \qquad \cdots (13)$$

Here, ti represents a tangential unit vector of the side

 l_i . When Expression (12) is substituted for Expression (13) and compared with Expression (10), the ij component P_{ij} of a prolongation matrix P is given by the following expression.

Expression 14

$$P_{ij} = \frac{1}{|I_i|} \int_{I_i} N_j^C(x) \cdot t_i dl \qquad \cdots (14)$$

The prolongation matrix forming portion 25 forms a prolongation matrix using Expression (14) and preserves the prolongation matrix in the prolongation matrix storage portion 27.

Next, the approximate value calculation portion 28 sets an approximate initial value v^F to u^F in order to calculate the electromagnetic field vector v^F using simultaneous linear equations according to a multi-grid method (S14). The initial value v^F may be zero. Moreover, the detaildetails of the calculation of the simultaneous linear equations according to the multi-grid method canmay be referred to in Reference Document 3 (W. Briggs, V. Henson and S. McCormick, "A Multigrid Tutorial," SIAM).

Next, the approximate value calculation portion 28 removes high-frequency components of error contained in v^F by applying a stationary iteration method of simultaneous linear equations, such as the Jacob method, the Gauss-Seidel method, or the SOR method, etc., to Expression (7) (S15). Moreover, the detaildetails of the iteration method can be referred to in Reference Document 4 (Satomi Hasegawa, Hidehiko Hasegawa, Kiyoji Fujino, "Iteration Method Templates", Asakura shotenShoten, 1996).

Next, the approximate value calculation portion 28 maps a residual r^F in Ω^F to a residual r^C in Ω^C by using the following

expression in order to remove low-frequency components (S16).

- Expression 13

Expressions 15 and 16

$$r^{F} = b^{F} - A^{F} v^{F} \qquad \cdots (15)$$

$$r^{C} = P^{t} r^{F} \qquad \cdots (16)$$

Next, the approximate value calculation portion 28 findsdetermines solutions of simultaneous linear equations in the following expression (17) and calculates a correction vector e^c (S17).

Expression
$$44\underline{17}$$

$$A^{C}e^{C}=r^{C} \qquad \cdots (17)$$

In Expression (17), although a non-stationary iteration method, such as a direct method according to the Gauss elimination method, a CG method, or a GMRES method, etc., is preferably used, since Expression (17) is a small matrix whenas compared with to Expression (7), the calculation can be performed in a short time. Moreover, the detaildetails of the iteration method can be referred to in the above-mentioned document 4.

Next, the approximate solution correction portion 29 improves the accuracy of the approximate solution by returning the correction vector of Ω^{c} to the correction vector of Ω^{f} by using the following Expressions (18) and (19) at (S18):

Expression 1518 and 19

$$e^F = Pe^C$$
 ... (18)
 $v^F = v^F + e^F$... (19)

The approximate solution correction portion 29 makes high-frequency components attenuated by applying smoothing shown in step 15 a couple of times to the approximate solution (S19).

Although the above processing in step S15 to S19 is the basic algorithm of a multi-grid method, since there are cases where the convergence is relatively slow, the convergence is quickened by accelerated by using another iteration method for simultaneous linear equations together as required (S20).

ConcretelySpecifically, the above-described processing is used as a pretreatment for a non-stationary iteration method, such as a CG method, a GMRES method, or a GCR method, etc., or the processing is accelerated by a residual cutting method, etc.

Moreover, the detaildetails of the residual cutting method can be referred to document in Reference Document 5 (Kikuchi et al., JSME International Journal, 62-604, Series B p. 4076—,

Next, the approximate solution correction portion 29 evaluates the residual of Expression (15) in step (S21). When the residual is large (S21 is no), since the electromagnetic field vector does not converge, the process $\frac{1}{3}$ $\frac{1}$

Fig. 8 shows one example of a rectangular waveguide as an analysis object of an electromagnetic field analyzer according to $\frac{\partial}{\partial x}$ preferred embodiment of the present invention. The

dimensions of this rectangular waveguide <u>isare</u> as shown in Fig. 8 and the inside is filled with air. The electromagnetic field analysis is performed using tetrahedral primary elements.

In Fig. 9, the calculation time to the number of elements when the rectangular waveguide shown in Fig. 8 is driven in TE10 mode of 2.45 GHz is compared between the case where a related direct method is used and the case where a method of according to the present invention is used. As is understood from Fig. 9, when the number of elements increases, the effect of the case where an electromagnetic field analyzer of according to the present preferred embodiment is used is eminentmore pronounced and, when the number of elements is 206,266, the speed of the processing is about 17 times as high as when compared with greater than that of the related direct method.

Fig. 10 shows one example of a patch antenna as an analysis object of the electromagnetic field analyzer according to ana preferred embodiment of the present invention. The dimensions of the patch antenna are as—shown in Fig. 10, and the patch antenna is formeddisposed on a dielectric substrate having a thickness of about 0.794 mm and a dielectric constant of about 2.2. The electromagnetic field analysis is performed by using primary tetrahedral elements.

In Fig. 11, the calculation time to the number of elements when the patch antenna shown in Fig. 10 is driven at 7.0 GHz is compared between the case where a related direct method is used and the case where a method according to a preferred embodiment of the present invention is used. As is understood from Fig. 11, when the number of elements increases, the

effect of the case where an electromagnetic field analyzer of the present <u>preferred</u> embodiment is used is <u>eminentpronounced</u> and, when the number of elements is 88,445, the speed of the processing is about 15 times as high as when compared with the than that of the related direct method.

As described above, according to the electromagnetic field analyzer of the present <u>preferred</u> embodiment, since a prolongation matrix P <u>makingdefining</u> the electromagnetic field vector \mathbf{u}^{F} in Ω^{F} related to the electromagnetic field vector \mathbf{u}^{C} in Ω^{C} is formed and an electromagnetic field vector is calculated by using this prolongation matrix, it <u>became is</u> possible to perform an electromagnetic field analysis by using a multi-grid method using a non-nested <u>mech</u>mesh.

Furthermore, since it became is possible to perform an electromagnetic field analysis by using a multi-grid method using a non-nested mesh, the time required for analysis can-be is greatly reduced whenas compared with to an electromagnetic field analysis using a direct method.

The embodiments described this time are exemplification in all aspects and not restrictive. The scope of the present invention is not shown by the above description, but shown by the scope of the claims, and the range equivalent to the scope of the claims and all modifications within the scope of the claims are to be included.

While the present invention has been described with respect to preferred embodiments, it will be apparent to those skilled in the art that the disclosed invention may be modified in numerous ways and may assume many embodiments other than those specifically set out and described above. Accordingly, it is intended by the appended claims to cover all modifications of
the present invention that fall within the true spirit and
scope of the invention.